



הרצאת העשרה בנושא  
**VHDL**  
(or beyond LAB #1 experiments)  
part II

Version 1.0 - May 2012

Amos Zaslavsky

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

Page: 1



Introduction

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

Page: 2



## VHDL-2008 is here (alive and kicking..)

- The **Original VHDL-1987** standard was very **powerful** !
- Previous updated revisions were minor
  - **VHDL-1993** (files, direct instantiation, extended naming of identifiers, some minor coding style change, shared variables .. )
  - **VHDL-2002** (protected shared variables, improved buffer mode, bigger real .. )
- Because the basic version was very powerful and modifications were very minor, many tool vendors ignored modifications or upgraded their tools very slowly (why bother ..)
- Many people consider only **Verilog 2001** to be a matching competitor to **VHDL** (this tie situation continues from 2001 until 2005)
- SystemVerilog (2005) is a superset of Verilog that improves its verification capabilities. Some are beyond VHDL's verification capabilities.
- **VHDL-2008** is a Major Revision with many improvement

A Short History  
Lesson

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 01

Page: 3



## Improvements in VHDL-2008 are in areas:

- Making VHDL language stronger (many more features.. )
- Making VHDL language easier to use (less cumbersome.. )
- Enhancements in the area of Verification (adding powerful features to existing capabilities + More coming soon.. )

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 01

Page: 4



## The final goal (מה מתבשל בתנור)

- Why update VHDL ? Instead, why not adopt SystemVerilog as verification Language ?
- VHDL already includes many system-level modeling features such as: records, access types (pointers), Protected types & shared variables, PSL and many many other advanced features that can be enhanced with relatively little effort, and **new features can be added in a way that integrates cleanly with existing features.**
- Organizations using VHDL already have significant experience using the language and a pool of people familiar with the language.
- Using SystemVerilog would force users to learn a language that is different or have to manage multilingual environments (causing lower productivity)
- Final Goal: **VHDL = Verification & Hardware Description Language**
- The next enhancements are mostly Verification focused:
  - Object Oriented Classes
  - Verification Data structures
  - More powerful Randomization (enhancing existing capabilities)
  - Functional Coverage
  - & Incorporating good things from SystemVerilog, SystemC, Vera & E



## Presentation goals

### מטרה

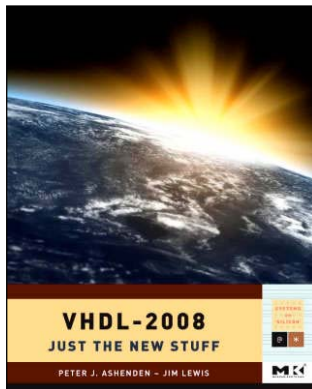
## ההצבעה על כיוונים

(מבלי להעמיק בחומר ומבלי לעבור על כל התוספות)



## We will no go into all the modifications..

- For More information ( or use the new LRM :-(-)



## Topics can all be found in



## Topics that will be Mentioned here

- More reserved words
- More convenient Comments
- More powerful & convenient constant string assignments
- More convenient mixed bit & vector Logical operations
- Unary reduction operations
- More built in arrays
- numeric\_std\_unsigned from IEEE replaces std\_logic\_unsigned from Synopsys
- Better relations
- New maximum(\_), & minimum(\_) functions
- New Fixed point and float vectors
- The Context design unit
- The condition operator
- More convenient Case operations
- conditional assignments & selected assignments in process

Simulation/Verification

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

Page: 9



## Topics that will be Mentioned here

- More convenient combinatorial descriptions
- More convenient sync descriptions
- New to\_string(\_) functions
- expressions written to ports when instantiation of components
- Revision of Modes (port directions)
- More convenient & powerful generate statements
- Forcing signals through hierarchies using external names
- Forcing signals through hierarchies using external names
- Forcing ports: Driving value & Effective value
- Embedded PSL
- IP (Intellectual Property) protection
- VHDL Procedural Interface (VHPI)

Simulation/Verification  
only

Simulation/Verification  
directed

Simulation/Verification  
only

Simulation/Verification  
directed

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

Page: 10



More reserved  
words

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 11



## Reserved names VHDL87

abs	access	after	alias	all
and	architecture	array	assert	attribute
begin	block	body	buffer	bus
case	component	configuration	constant	disconnect
downto	else	elsif	end	entity
exit	file	for	function	generate
generic	guarded	if	in	inout
is	label	library	linkage	loop
map	mod	nand	new	next
nor	not	null	of	on
open	or	others	out	package
port	procedure	process	range	record
register	rem	report	return	select
severity	signal	subtype	then	to
transport	type	units	until	use
variable	wait	when	while	with
xor				

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 12



## More reserved names VHDL-93/2002/2008

group	impure	inertial	literal	posponed
pure	reject	rol	ror	shared
sla	sll	sra	srl	unaffected
xnor				

VHDL-93

protected

VHDL-2002

VHDL-2008

assume	assume_guarantee	context	cover
default	fairness	force	parameter
property	release	restrict	restrict_guarantee
sequence	strong	vmode	vprop
vunit			

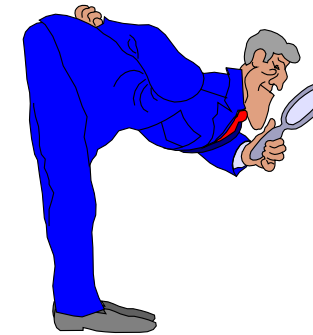
PSL words

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 13



More convenient  
Comments

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 14



## Comments

```
-- this comment ends with <CR> <LF>
```

```
y <= a and b and c ; -- right part of this line is a comment
```

```
-- this  
-- is  
-- a  
-- long  
-- comment
```



```
/* this is a delimiter  
comment in VHDL-2008 */
```

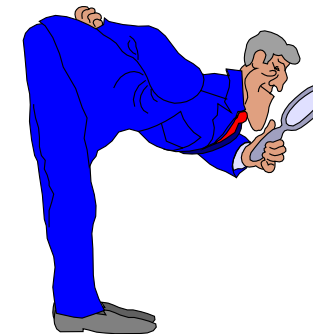


8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 15



More powerful &  
convenient constant  
string assignments

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 16



## Old & New bit-string literals

```

y <= b"11110110" ;           -- binary
y <= x"F6" ;                 -- hex
y <= "1111" & "011" & '0' ; -- default vector is bin
y <= x"f" & o"3" & '0' ;    -- octal & lowercase too
    
```

ok on older versions

```

y <= 246 ; -- error
y <= d"246" ; -- ERROR d is not a legal prefix
    
```



```

y <= d"246" ; -- OK in VHDL-2008 !
    
```



## Specifying width in new bit-string literals

### ■ Padding with leading zeros on the left

```

13d"246"    => "0000011110110"
13o"366"    => "0_000_011_110_110"
13x"f6"     => "0_0000_1111_0110"
10o"0366"   => "0_011_110_110"
10x"0f6"    => "00_1111_0110"
    
```

### ■ Signed values (Unsigned is the default)

```

12ux"f6" => "0000_1111_0110" (unsigned extension)
12x"f6"  => "0000_1111_0110" (unsigned extension)
12sx"f6" => "1111_1111_0110" (signed extension)
    
```

### ■ assigning std\_logic\_vector with meta-characters

```

12o"01X" => "000_111_XXX"
    
```



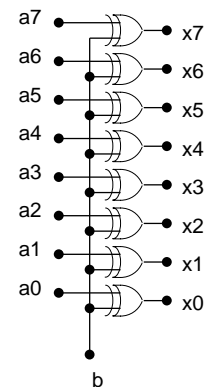
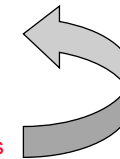
## Mixed vector & bit operations

-- a polarity changer ?

```

entity op_bv1 is
  port ( x : out bit_vector(7 downto 0) ;
        a : in bit_vector(7 downto 0) ;
        b : in bit
    ) ;
end op_bv1 ;
architecture arc_op_bv1 of op_bv1 is
begin
  x <= a xor b ;
end arc_op_bv1 ;
    
```

Not legal on older Versions



OK on VHDL-2008

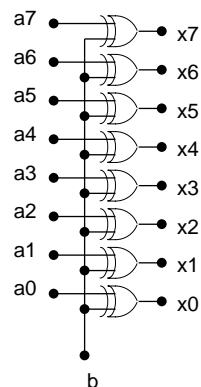


More convenient  
mixed bit & vector  
Logical operations



## Solution on older versions

```
-- a polarity changer
entity op_bv1 is
  port ( x : out bit_vector(7 downto 0) ;
        a : in  bit_vector(7 downto 0) ;
        b : in  bit
        ) ;
end op_bv1 ;
architecture arc_op_bv1 of op_bv1 is
  signal b_bv : bit_vector ( 7 downto 0 ) ;
begin
  b_bv <= ( others => b ) ;
  x <= a xor b_bv ;
end arc_op_bv1 ;
```



## Solution on older versions

```
-- some other solutions (less elegant)
```

```
x <= a xor ( b & b & b & b & b & b & b & b ) ;
x <= a xor ( b , b , b , b , b , b , b , b ) ;
x <= ( a(7) xor b , a(6) xor b , a(5) xor b , a(4) xor b ,
      a(3) xor b , a(2) xor b , a(1) xor b , a(0) xor b ) ;
x(7) <= a(7) xor b ;
x(6) <= a(6) xor b ;
x(5) <= a(5) xor b ;
x(4) <= a(4) xor b ;
.
..
```



## What had to be done in older versions

```
-- no reduced operations on old VHDL
```

```
entity reduced_vec is
  port ( din      : in  bit_vector(7 downto 0) ;
        parity   : out bit
        ) ;
end reduced_vec ;
architecture arc_reduced_vec of reduced_vec is
begin
  parity <= din(7) xor din(6)
          xor din(5) xor din(4)
          xor din(3) xor din(2)
          xor din(1) xor din(0);
end arc_reduced_vec ;
```

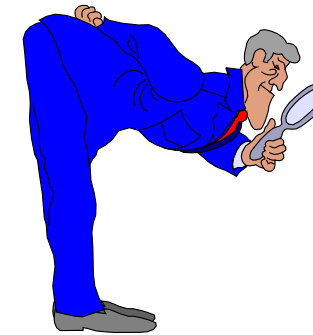


Unary reduction operations



## Unary reduction operations in VHDL-2008

```
-- reduced operations on a vector (VHDL-2008)
entity reduced_vec is
  port ( din      : in  bit_vector(7 downto 0) ;
        parity   : out bit
        ) ;
end reduced_vec ;
architecture arc_reduced_vec of reduced_vec is
begin
  parity <= xor din ;
end arc_reduced_vec ;
```



More built in arrays



## boolean, integer, real & time arrays

```
-- decoder 3->8 using numeric relations
entity dec3t8 is
  port ( din      : in  integer range 0 to 7 ;
        enable   : in  boolean
        ) ;
  d      : out  boolean_vector(7 downto 0) ;
end dec3t8 ;
architecture arc_dec3t8 of dec3t8 is
begin
  -- positional association
  d <= ( ( din=7 ) and enable ,
        ( din=6 ) and enable ,
        ( din=5 ) and enable ,
        ( din=4 ) and enable ,
        ( din=3 ) and enable ,
        ( din=2 ) and enable ,
        ( din=1 ) and enable ,
        ( din=0 ) and enable ) ;
end arc_dec3t8 ;
```

in old VHDL only bit\_vector  
(or std\_logic\_vector) and  
strings

boolean\_vector,  
integer\_vector, real\_vector,  
time\_vector



## boolean, integer, real & time arrays

```
-- decoder 3->8 using numeric relations
entity dec3t8 is
  port ( din      : in  integer range 0 to 7 ;
        enable   : in  boolean
        ) ;
  d      : out  boolean_vector(7 downto 0) ;
end dec3t8 ;
architecture arc_dec3t8 of dec3t8 is
begin
  -- named association
  d <= ( 7 => ( din=7 ) and enable ,
        6 => ( din=6 ) and enable ,
        5 => ( din=5 ) and enable ,
        4 => ( din=4 ) and enable ,
        3 => ( din=3 ) and enable ,
        2 => ( din=2 ) and enable ,
        1 => ( din=1 ) and enable ,
        0 => ( din=0 ) and enable ) ;
end arc_dec3t8 ;
```

in old VHDL only bit\_vector  
(or std\_logic\_vector) and  
strings

boolean\_vector,  
integer\_vector, real\_vector,  
time\_vector





`std_logic_unsigned` from  
Synopsys is replaced by  
`numeric_std_unsigned` from  
IEEE



## New: numeric\_std\_unsigned

- The most popular 3 declarations (math operations on vectors)
  - Using the classic package of Synopsys in old VHDL
  - `library ieee ;`
  - `use ieee.std_logic_1164.all ;`
  - `use ieee.std_logic_unsigned.all ;`
- Replaced by:
  - Same thing with new package of IEEE in VHDL-2008
  - `library ieee ;`
  - `use ieee.std_logic_1164.all ;`
  - `use ieee.numeric_std_unsigned.all ;` -- a new package !
- More operations (divisions,power)

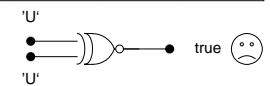
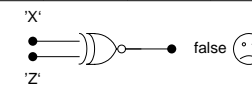
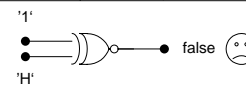


Better relations



## Problems with old = relations

=	'U'	'X'	'0'	'1'	'Z'	'W'	'L'	'H'	'-'
'U'	true	false	false	false	false	false	false	false	false
'X'	false	true	false	false	false	false	false	false	false
'0'	false	false	true	false	false	false	false	false	false
'1'	false	false	false	true	false	false	false	false	false
'Z'	false	false	false	false	true	false	false	false	false
'W'	false	false	false	false	false	true	false	false	false
'L'	false	false	false	false	false	false	true	false	false
'H'	false	false	false	false	false	false	false	true	false
'-'	false	false	false	false	false	false	false	false	true



Not behaving like real hardware  
Return Boolean





## New (matched) ?= relation

?=	'U'	'X'	'0'	'1'	'Z'	'W'	'L'	'H'	'-'
'U'	'U'	'X'	'0'	'1'	'Z'	'W'	'L'	'H'	'-'
'X'	'U'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'
'0'	'U'	'X'	'0'	'0'	'X'	'X'	'0'	'0'	'X'
'1'	'U'	'X'	'0'	'1'	'X'	'X'	'0'	'1'	'X'
'Z'	'U'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'
'W'	'U'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'
'L'	'U'	'X'	'1'	'0'	'X'	'X'	'1'	'0'	'X'
'H'	'U'	'X'	'0'	'1'	'X'	'X'	'0'	'1'	'X'
'-'	'1'	'1'	'1'	'1'	'1'	'1'	'1'	'1'	'1'



Vectors are and-ed together, with any 'X' producing 'X' result and any '0' producing '0' result

Behaving like real hardware



## Problems with old relations - (<)

<	right								
left	'U'	'X'	'0'	'1'	'Z'	'W'	'L'	'H'	'-'
'U'	false	true	true	true	true	true	true	true	true
'X'	false	false	true	true	true	true	true	true	true
'0'	false	false	false	true	true	true	true	true	true
'1'	false	false	false	false	true	true	true	true	true
'Z'	false	false	false	false	false	true	true	true	true
'W'	false	false	false	false	false	false	true	true	true
'L'	false	false	false	false	false	false	false	true	true
'H'	false	false	false	false	false	false	false	false	true
'-'	false	false	false	false	false	false	false	false	false

'U' < 'X' < '0' < '1' < 'Z' < 'W' < 'L' < 'H' < '-'

No Meaningful hardware behavior

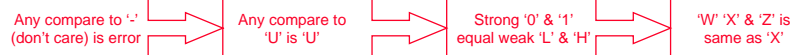


Return Boolean



## New (matched) ?< relation

?<	right								
left	'U'	'X'	'0'	'1'	'Z'	'W'	'L'	'H'	'-'
'U'	'U'	'X'	'0'	'1'	'Z'	'W'	'L'	'H'	'-'
'X'	'U'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'
'0'	'U'	'X'	'0'	'0'	'X'	'X'	'0'	'0'	'X'
'1'	'U'	'X'	'0'	'1'	'X'	'X'	'0'	'1'	'X'
'Z'	'U'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'
'W'	'U'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'
'L'	'U'	'X'	'0'	'1'	'X'	'X'	'0'	'1'	'X'
'H'	'U'	'X'	'0'	'0'	'X'	'X'	'0'	'0'	'X'
'-'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'	'X'



Behaving like real hardware



## Other Matched relations

- Are all created by operations on the previous relations

?/= is an inverted ?=

?>= is inverted ?<

?<= is ?< or-ed with ?=

- Example of usage:

```
library ieee ;
use ieee.std_logic_1164.all ;
use ieee.numeric_std_unsigned.all ; -- present for math meaning
entity good_compare2 is
  port ( a      : in std_logic_vector(3 downto 0) ;
        b      : in std_logic_vector(2 downto 0) ;
        a_gte_b : out std_logic ) ;
end good_compare2 ;
architecture arc_good_compare2 of good_compare2 is
begin
  a_gt_b <= ( a ?>= b ) ; -- matching directional compare
end arc_good_compare2 ;
```





## Order of precedence of operators



## Order of Precedence of operators

operator level (low to high)	Operators
condition	??
logical	and, or, nand, nor, xor, xnor
relational	=, /=, <, <=, >, >=, ?=, ?/=, ?<, ?<=, ?>, ?>=
shift	sll, slr, sla, sra, rol, ror
adding	+, -, &
sign	+, -
multiplying	*, /, mod, rem
miscellaneous	** , abs, not
parentheses	()



## New maximum(\_), & minimum(\_) functions



## minimum(\_) & maximum(\_) functions

- Can be used with the following data types: `character`, `bit`, `integer`, `real`, `time`, `string`, `boolean_vector`, `bit_vector`, `integer_vector`, `real_vector`, `time_vector`, `std_logic_vector` and many more types .. (any data types that relations can be operated on)

- Usage:

```
maximum( object1 , object2 ) -- max between 2
```

```
maximum ( array_type ) -- returns max array element
```

- Example of usage:

```
constant w1 : integer := 3 ;
```

```
constant w2 : integer := 2 ;
```

```
signal y : std_logic_vector(maximum(w1,w2) downto 0) ;
```





New Fixed point and float vectors



## More Powerful vector data types

- The new Powerful data types
  - Unsigned fixed - *ufixed*
  - Signed fixed - *sfixed*
  - Floating Point - *float*
- They are all vectoric data types !
- Useful for:
  - Large dynamic range (including fractions)
  - Complex Arithmetic and DSP applications
- fixed types consume less hardware resources but have lower dynamic range, float has more dynamic range but hardware is more complex
- Support for synthesis: In a short time
  - For example, Altera has already the flowing floating point macrofunctions: *ALTFP\_ABS*, *ALTFP\_ADD\_SUB*, *ALTFP\_COMPARE*, *ALTFP\_CONVERT*, *ALTFP\_DIV*, *ALTFP\_EXP*, *ALTFP\_INV*, *ALTFP\_INV\_SQRT*, *ALTFP\_LOG*, *ALTFP\_MULT*, *ALTFP\_SQRT*, and more coming soon... & fixed = integers



## ufixed & sfixed declarations & assignments

- Declarations & assignments (negative index range is fraction)
 

```
signal u : ufixed (1 downto -4) ;
signal s : sfixed (3 downto -1) ;
...
u <= "101010" ; -- 10.1010 = 2.625
s <= "11100" ; -- 1110.0 = -2
```
- Equivalent Assignments (parameters are index ranges)
 

```
u <= to_ufixed(2.625,1,-4) ; -- 10.1010 = 2.625
s <= to_sfixed( -2 ,3,-1) ; -- 1110.0 = -2
```
- Equivalent Assignments (parameter copies ranges from declarations of u and s respectively)
 

```
u <= to_ufixed(2.625,u) ; -- 10.1010 = 2.625
s <= to_sfixed( -2 ,s) ; -- 1110.0 = -2
```



## ufixed & sfixed declarations & assignments

- Code example of an adder
 

```
library ieee ;
use ieee.fixed_pkg.all ;
entity fixed_vec is
  port ( a,b : in ufixed(3 downto -4) ; -- width 8
        y : out ufixed(4 downto -4) ) ; -- width 9
end fixed_vec ;
architecture arc_fixed_vec of fixed_vec is
begin
  y <= a + b ; -- both sides are width 9
end arc_fixed_vec ;
```
- Operands are 8 bit wide & Result is 9 bits (Full Precision math with no automatic modulo-truncation)

Example

c	c	c	c				
1	1	1	1	.	0	0	0
1	1	1	1	.	0	0	0
1	1	1	1	0	.	0	0



## with old vector (integer) math

- Both sides of assignments are not equal so it is not a legal code

```
-- error
library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_unsigned.all ;
entity vec_math is
  port ( a,b : in std_logic_vector(3 downto 0) ;
        y : out std_logic_vector(4 downto 0) ) ;
end vec_math ;
architecture arc_vec_math of vec_math is
begin
  y <= a + b ; -- error
end arc_vec_math ;
```



## with old vector (integer) math

- Now you get automatic modulo-truncation

```
-- OK but with different results
library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_unsigned.all ;
entity vec2_math is
  port ( a,b : in std_logic_vector(3 downto 0) ;
        y : out std_logic_vector(3 downto 0) ) ;
end vec2_math ;
architecture arc_vec2_math of vec2_math is
begin
  y <= a + b ;
end arc_vec2_math ;
```

Example

<del>c</del>	c	c	c
1	1	1	1
1	1	1	1
1	1	1	0



## with old vector (integer) math

- Handling the truncated bit

```
-- Handling truncation bit
library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_unsigned.all ;
entity vec3_math is
  port ( a,b : in std_logic_vector(3 downto 0) ;
        y : out std_logic_vector(4 downto 0) ) ;
end vec3_math ;
architecture arc_vec3_math of vec3_math is
begin
  y <= ('0' & a) + ('0' & b) ;
end arc_vec3_math ;
```



## Other operations on ufixed

- Result size

operation	result range
A + B , A - B	Max(A'left,B'left)+1 downto Min(A'right,B'right)
A * B	A'left+B'left+1 downto A'right+B'right
A / B	A'left-B'right downto A'right-B'left-1
A rem B	Min(A'left,B'left) downto Min(A'right,B'right)
A mod B	B'left downto Min(A'right,B'right)

- example:

```
signal A : ufixed (7 downto -4) ;
signal B : ufixed (3 downto -9) ;
```

- Width of multiplication result

```
ufixed(7+3+1 downto (-4)+(-9))
```

- No meaning for `abs(_)` and unary `-(_)` with `ufixed`



## Other operations on sfixed

- Result size

operation	result range
A + B , A - B	Max(A'left,B'left)+1 downto Min(A'right,B'right)
A * B	A'left+B'eft+1 downto A'right + B'right
A / B	A'left-B'right+1 downto A'right-B'left
A rem B	Min(A'left,B'left) downto Min(A'right,B'right)
A mod B	Min(A'left,B'left) downto Min(A'right,B'right)
abs(A)	A'left+1 downto A'right
-A	A'left+1 downto A'right

- Added operations meaning for `abs(_)` and unary `-(_)` with `sfixed`



## More operations

- Mixing with integer and real

```
signal a : sfixed (8 downto -5);
```

```
signal b : real ;
```

```
...
```

```
y <= a + b ;
```

- Equivalent to last assignment (b converted to ranges of a)

```
y <= a + sfixed(b,8,-5) ;
```

- Relation (Compare operations)

```
= , /= , < , > , >= , <= , ?= , ?/= , ?< , ?> , ?>= , ?<=
```

- Logical operations

```
and, or, nand, nor, xnor, not
```

- Conversion functions

```
sll, srl, rol, ror, sla, sra
```

- Conversion functions

```
to_sfixed(_),to_ufixed(_),to_real(_),to_integer(_),to_std_logic(_), to_string(_),resize(_)
```



## Beyond fixed - float

- Float ⇔ Fixed

- more (dynamic range + hardware) ⇔ less (dynamic range + hardware)
- constant relative accuracy ⇔ constant absolute accuracy

- What's wrong with real

- Fixed range of 64 bit not suitable for all applications
- No easy and direct access to bits of number

- Package `float_pkg` enables usage of

- Standard 32 bit & 64 bit numbers (IEEE-754)
- Standard 128 bit numbers
- Any width (compliant with IEEE-854)

- 3 Parts (mandatory)

```
sign.exponent.fraction
```

- Examples in next foils..

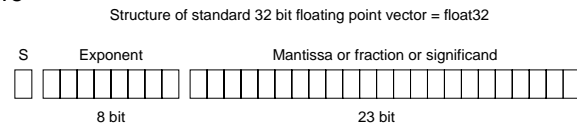


New Floating point  
vectors



## float32 example (normalized)

- Structure



- Normalization method of fraction is 1.fraction (1 not represented)
- Calculation of a normalized number

$$NORMALIZED\_NUM = (-1)^{SIGN} \cdot 2^{EXONENT-127} \cdot (1.0 + FRACTION)$$

- example:

```
use ieee.float_pkg.all ;
..
signal f : float (8 downto -23) ; -- bit 8 is sign
..
begin
f <= to_float(6.5, 8, 23) ;
```



## float32 example (normalized)

- the previous function `to_float(_)` converts real to float
- Bit 8 (MSB) is the sign bit
- bits (7 `downto` 0) are for exponent
- bits (-1 `downto` 23) are for mantissa=fraction=significand
- Direction of vector must always be `downto`

- Equivalent declaration to the previous one:

```
signal f : float32 ; -- a shorter way
```

- Equivalent assignments:

```
f <= to_float(6.5, f) ;
f <= '0' & "10000001" & "101000000000000000000000" ;
```

- Calculation of last assignment:

$$\begin{aligned} & (-1)^{SIGN} \cdot 2^{EXONENT-127} \cdot (1.0 + FRACTION) \\ &= 2^{129-127} \cdot (1.0 + 0.625) = (4) \cdot (1.625) = 6.5 \end{aligned}$$



## float32 example (normalized)

- The smallest possible normalized values (exponent must not be zero)

```
f <= '0' & "00000001" & "000000000000000000000000" ;
f <= '1' & "00000001" & "000000000000000000000000" ;
```

- Calculation of the smallest normalized positive value

$$\begin{aligned} & (-1)^{SIGN} \cdot 2^{EXONENT-127} \cdot (1.0 + FRACTION) \\ &= 2^{1-127} \cdot (1.0 + 0.0) = 2^{-126} \\ &\approx (1.175494350) \cdot 10^{-38} \end{aligned}$$

- exponent 00000000 is considered un-normalized and is calculated differently (next foils)



## float32 example (un-normalized)

- Un-normalized (De-Normalized) values have a zero exponent

$$DENORMALIZED\_NUMBER = (-1)^{SIGN} \cdot 2^{-126} \cdot (FRACTION)$$

- Biggest values are & calculation of biggest positive value:

```
f <= '0' & "00000000" & "111111111111111111111111" ;
f <= '1' & "00000000" & "111111111111111111111111" ;
```

$$\begin{aligned} & DENORMALIZED\_NUMBER = (-1)^{SIGN} \cdot 2^{-126} \cdot (FRACTION) \\ &= (-1)^{SIGN} \cdot 2^{-126} \cdot (1 - 2^{-23}) \\ &\approx (1.175494210) \cdot 10^{-38} \end{aligned}$$

- Mid values are & calculation of mid positive value:

```
f <= '0' & "00000000" & "100000000000000000000000" ;
f <= '1' & "00000000" & "100000000000000000000000" ;
```

$$\begin{aligned} & DENORMALIZED\_NUMBER = (-1)^{SIGN} \cdot 2^{-126} \cdot (FRACTION) \\ &= 2^{-126} \cdot (0.5) = 2^{-127} \\ &\approx (5.87747..) \cdot 10^{-39} \end{aligned}$$



## float32 example (un-normalized)

- Smallest values are & calculation smallest positive value:

```
f <= '0' & "00000000" & "000000000000000000000001" ;
f <= '1' & "00000000" & "000000000000000000000001" ;
```

$$\begin{aligned} \text{DENORMALIZED\_NUMBER} &= (-1)^{\text{SIGN}} \cdot 2^{-126} \cdot (\text{FRACTION}) \\ &= 2^{-126} \cdot (2^{-23}) = 2^{-149} \\ &\approx (1.401298...) \cdot 10^{-45} \end{aligned}$$

- Any smaller number than that, is considered zero & calculation of it's positive value (zero) is:

```
f <= '0' & "00000000" & "000000000000000000000000" ;
f <= '1' & "00000000" & "000000000000000000000000" ;
```

$$\begin{aligned} \text{DENORMALIZED\_NUMBER} &= (-1)^{\text{SIGN}} \cdot 2^{-126} \cdot (\text{FRACTION}) \\ &= (-1)^{\text{SIGN}} \cdot 2^{-126} \cdot (0) \\ &= 0 \end{aligned}$$



## float32 example (big-normalized)

- The biggest possible normalized values (exponent no zeros) & calculation of it's value

```
f <= '0' & "11111110" & "111111111111111111111111" ;
f <= '1' & "11111110" & "111111111111111111111111" ;
```

$$\begin{aligned} &(-1)^{\text{SIGN}} \cdot 2^{\text{EXPONENT}-127} \cdot (1.0 + \text{FRACTION}) \\ &= 2^{254-127} \cdot (1.0 + (1 - 2^{-23})) \\ &= 2^{+127} \cdot (2.0 - 2^{-23}) \\ &\approx (3.402823...) \cdot 10^{38} \end{aligned}$$

- Values beyond that are considered infinite:

```
f <= '0' & "11111111" & "000000000000000000000000" ;
f <= '0' & "11111111" & "000000000000000000000000" ;
```

- Infinite is result of division of non zero by zero.



## float32 example (NaN)

- Exponent 11111111 and fraction that is not all zeros is NaN

- Result of:

- division of 0 by 0
- Square root of -1
- Addition of +infinity and -infinity
- division of infinity by infinity
- others..

- The fraction is considered a Payload that specifies code of problem

- Quiet NaN or qNaN (MSB of fraction is 1) ⇔ Signaling NaN or sNaN (MSB of fraction is 0)

```
f <= '0' & "11111110" & "1????????????????????????" ;
f <= '1' & "11111110" & "0????????????????????????" ;
```

- After Quiet NaN or qNaN calculations can continue

- After Signaling NaN or sNaN calculations are stopped



## Example of floating point divider

```
-- example of a floating point divider
```

```
library ieee ;
```

```
use ieee.float_pkg.all ;
```

```
entity fdiv is
```

```
    port ( a,b : in float32 ;
           y   : out float32 ) ;
```

```
end fdiv ;
```

```
architecture arc_fdiv of fdiv is
```

```
begin
```

```
    y <= a / b ;
```

```
end arc_fdiv ;
```

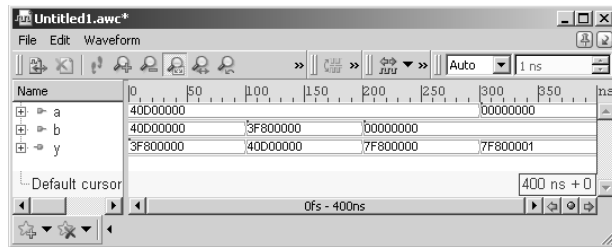
Non Standard floating points can be declared too

```
standard 64 & 128 bit
signal a,b : float64 ;
signal a,b : float128 ;
-- non standard 16 bit
signal a,b : float(5 downto -10) ;
-- non standard 7 bits (minimum)
signal a,b : float(3 downto -3) ;
```



## Example of floating point divider

```
force a x"40d00000" ;# 6.5
force b x"40d00000" ;# 6.5
run 100 ns ;# expected x3f800000 = 1.0
force b x"3f800000" ;# 1.0
run 100 ns ;# expected x40d00000 = 6.5
force b x"00000000" ;# 0.0
run 100 ns ;# expected x7f800000 = inf
force a x"00000000" ;# 0.0
run 100 ns ;# expected x7f800001 = NaN
```



8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 61



## More standard float formats (64 & 128 bits)

- Declaring standard 64 & 128 bit standard floats
 

```
signal f1 : float(11 downto -52) ; -- standard 64
signal f2 : float(15 downto -112) ; -- standard 128
```
- An alternative declarations
 

```
signal f1 : float64 ; -- standard 64 bit
signal f2 : float128 ; -- standard 128 bit
```
- The general Normalized formula:

$$NORMALIZED\_NUM = (-1)^{SIGN} \cdot 2^{EXPONENT - (EXPONENT\_BIAS)} \cdot (1.0 + FRACTION)$$

$$EXPONENT\_BIAS = 2^{EXPONENT\_WIDTH - 1} - 1$$

- The General De-Normalized Formula:

$$DENORMALIZED\_NUM = (-1)^{SIGN} \cdot 2^{-(EXPONENT\_BIAS - 1)} \cdot (FRACTION)$$

$$EXPONENT\_BIAS = 2^{EXPONENT\_WIDTH - 1} - 1$$

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 62



## More standard float formats (64 & 128 bits)

- The Standard 64 float Formula:

$$NORMALIZED\_N_{64} = (-1)^{SIGN} \cdot 2^{EXPONENT - 1023} \cdot (1.0 + FRACTION)$$

$$DENORMALIZED\_N_{64} = (-1)^{SIGN} \cdot 2^{-(1022)} \cdot (FRACTION)$$

- The Standard 128 float Formula:

$$NORMALIZED\_N_{128} = (-1)^{SIGN} \cdot 2^{EXPONENT - 16383} \cdot (1.0 + FRACTION)$$

$$DENORMALIZED\_N_{128} = (-1)^{SIGN} \cdot 2^{-(16382)} \cdot (FRACTION)$$

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 63



## More on float format

- Non standards of 16 bits & 7 bits examples or others possible too..
 

```
signal f3 : float(5 downto -10) ; -- 16 bit
signal f4 : float(3 downto -3) ; -- minimum 7 bits
```
- Operations supported
 

```
+, -, *, /, rem, mod, abs(_), -(unary)
```
- Ranges are adjusted to the widest exponent & fraction

```
-- adjusting floats with different ranges
library ieee ; use ieee.float_pkg.all ;
entity fadd is
port ( a : in float( 8 downto -23) ;
      b : in float(11 downto -52) ;
      y : out float(11 downto -52) ) ;
end fadd ;
architecture arc_fadd of fadd is
begin
    y <= a + b ;
end arc_fadd ;
```

Adjusted to this

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 02

Page: 64





## More on float format

- Last assignment is equivalent to:

```
y <= resize(a,11,52) + b ;
```

```
y <= resize(a,b) + b ;
```

- type `real` is automatically adjusted to the float operand:

```
-- real type adjusted to float
```

```
library ieee ; use ieee.float_pkg.all ;
```

```
entity fadd2 is
```

```
    port ( a : in float(8 downto -23) ;
```

```
          b : in real ;
```

```
          y : out float(8 downto -23) ) ;
```

Adjusted to this

```
end fadd2 ;
```

```
architecture arc_fadd2 of fadd2 is
```

```
begin
```

```
    y <= a + b ;
```

```
end arc_fadd2 ;
```



## More on float format

- Last assignment is equivalent to:

```
y <= a + to_float(b,a) ;
```

- If any operand has non strong '0' or '1' or non strong 'L' or 'H' result is "XX..X"

- Relations are all supported

```
= , /= , < , > , >= , <=
```

```
?= , ?/= , ?< , ?> , ?>= , ?<=
```

- Logical operations are all supported

```
and, or, nand, nor, xnor, not
```

- Many functions supported:

```
- to_float(_), to_real(_), to_integer(_), to_stdlogic(_), to_stdlogic_vector(_)
```

```
- resize(_)
```

```
- to_string(_)
```

```
- more...
```

- Will soon be supported by synthesis !



The Context  
design unit



## The context design unit

- Makes including a group of libraries and packages easier using a single name reference

- Example of declaration (and then declaration to library `proj_lib`):

```
context easy is
    library ieee ;
    use ieee.std_logic_1164.all ;
    use ieee.std_logic_unsigned.all ;
    library proj_lib ;
    use proj_lib.proj_pack.all ;
    use std.textio.all ;
    use ieee.std_logic_textio.all ;
end context ;
```

- Example of usage:

```
library proj_lib ;
context proj_lib.easy ;
entity . . . is
```

```
. . .
```





The condition operator



## The ?? condition Operator

```
-- Introducing the condition operator "??"  
library ieee ;  
use ieee.std_logic_1164.all ;  
entity condition is  
    port ( din : in std_logic ;  
          dout : out boolean ) ;  
end condition ;  
architecture arc_condition of condition is  
begin  
    dout <= ?? din ; -- convert std_logic to boolean  
end arc_condition ;
```

- '1' or 'H' is converted to true
- other values ('0','L','X','Z','U','W','-') are converted to false



## Old style (not convenient)

```
entity condition1 is  
    port( cs0,cs1,pol : in bit ;  
          din : in bit_vector(3 downto 0) ;  
          dout : out bit_vector(3 downto 0) ) ;  
end condition1 ;  
architecture arc_condition1 of condition1 is  
begin  
    process (din,cs0,cs1,pol)  
    begin  
        if cs1 = '1' and cs0 = '0' and pol = '1' then  
            dout <= din ;  
        elsif cs1 = '1' and cs0 = '0' and pol = '0' then  
            dout <= not din ;  
        else  
            dout <= ( others => '0' ) ;  
        end if ;  
    end process ;  
end arc_condition1 ;
```



## Old style (using the ?? operator)

```
-- using the ?? operator  
architecture arc_condition2 of condition2 is  
begin  
    process (din,cs0,cs1,pol)  
    begin  
        if ?? ( cs1 and not cs0 and pol ) then  
            dout <= din ;  
        elsif ?? ( cs1 and not cs0 and not pol ) then  
            dout <= not din ;  
        else  
            dout <= ( others => '0' ) ;  
        end if ;  
    end process ;  
end arc_condition2 ;
```



## Old style (?? implicitly implied)

```
-- the ?? operator implicitly implied
architecture arc_condition3 of condition3 is
begin
  process (din,cs0,cs1,pol)
  begin
    if   cs1 and not cs0 and   pol then
      dout <= din ;
    elsif cs1 and not cs0 and not pol then
      dout <= not din ;
    else
      dout <= ( others => '0' ) ;
    end if ;
  end process ;
end arc_condition3 ;
```

OK too



## Where can this **implicitly implied ??** be used

- After an if clause (as in previous example) in an if statement
- After an elsif clause in an if statement
- After an until clause in a wait statement
- After an assert statement
- After a while clause in a while loop statement
- After a when clause in a next statement
- After a when clause in an exit statement
- After an if clause (as in previous example) in an if generate statement
- After an elsif clause in an if generate statement
- In a boolean expression in a PSL declaration
- In a boolean expression in a PSL statement



## There must be no ambiguity allowed

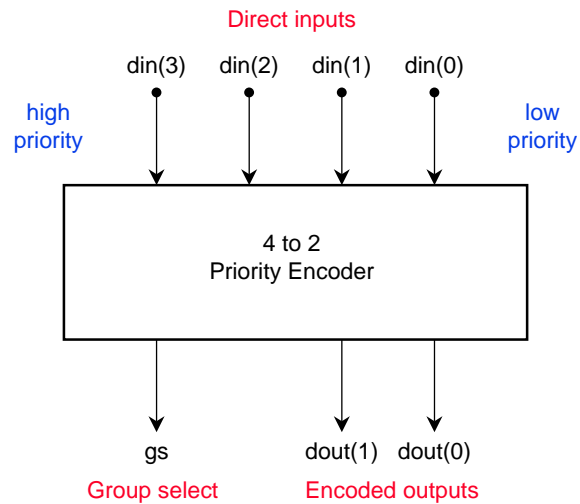
```
-- implicit conversion NOT applied
-- because of ambiguous interpretation !
architecture arc_condition4 of condition4 is
begin
  process (din,cs0,cs1,pol)
  begin
    if   cs1 and not cs0 and pol = '1' then -- error
      dout <= din ;
    elsif cs1 and not cs0 and pol = '0' then -- error
      dout <= not din ;
    else
      dout <= ( others => '0' ) ;
    end if ;
  end process ;
end arc_condition4 ;
```



More convenient  
Case operations



## Priority encoder example (again)



## The old case statement

```

case to_x01(din) is
  when "1000" | "1001" | "1010" | "1011"
    | "1100" | "1101" | "1110" | "1111" =>
    dout <= "11" ; gs <= '1' ;
  when "0100" | "0101" | "0110" | "0111" =>
    dout <= "10" ; gs <= '1' ;
  when "0010" | "0011" =>
    dout <= "01" ; gs <= '1' ;
  when "0001" =>
    dout <= "00" ; gs <= '1' ;
  when "0000" =>
    dout <= "00" ; gs <= '0' ;
  when others =>
    dout <= "XX" ; gs <= 'X' ; report "unknown din" ;
end case ;
    
```



## An new elegant matching `case?` statement

```

case? din is -- using the ?= (built in matching) operator
  when "1---" <=>
    dout <= "11" ; gs <= '1' ;
  when "01--" <=>
    dout <= "10" ; gs <= '1' ;
  when "001-" =>
    dout <= "01" ; gs <= '1' ;
  when "0001" =>
    dout <= "00" ; gs <= '1' ;
  when "0000" =>
    dout <= "00" ; gs <= '0' ;
  when others =>
    dout <= "XX" ; gs <= 'X' ; report "unknown din" ;
end case ;
    
```

'1' in code is treated as  $\Phi$

'H' is treated as '1'  
'L' is treated as '0'  
no need for to\_xo1()



## Works with selected assignments too

```

-- direct priority filter with
-- matching selected assignment
library ieee ;
use ieee.std_logic_1164.all ;
entity priority is
  port ( din : in std_logic_vector(3 downto 0) ;
        dout : out std_logic_vector(3 downto 0) ) ;
end priority ;
architecture arc_priority of priority is
begin
  with din select? -- VHDL-2008
    dout <= "1000" when "1---" ,
          "0100" when "01--" ,
          "0010" when "001-" ,
          "0001" when "0001" ,
          "0000" when others ;
end arc_priority ;
    
```



## Complex assignments in process (1 of 2)



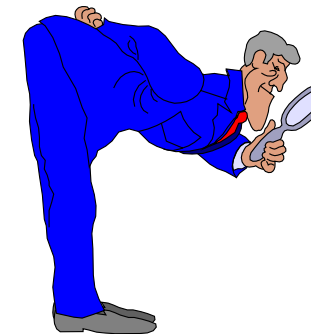
conditional  
assignments &  
selected  
assignments in  
process

```
-- mixed 4=>1 mux with enable
entity mixedmux is
    port ( din : in bit_vector(3 downto 0) ;
          sel : in bit_vector(1 downto 0) ;
          ena : in bit ; -- ena
          dout : out bit
        ) ;
end mixedmux ;
architecture arc_mixedmux of mixedmux is
begin
    process (din,sel,ena)
```



## Complex assignments in process (2 of 2)

```
process (din,sel,ena)
begin
    if sel = "00" then
        -- conditional assignment in process
        dout <= din(0) when ena = '1' else '0' ;
    elsif sel = "01" then
        -- conditional assignment in process
        dout <= din(1) when ena = '1' else '0' ;
    elsif sel = "10" then
        -- selected assignment in process
        with ena select
            dout <= din(2) when '0' ,
                  '0' when others ;
    else -- sel="11"
        -- selected assignment in process
        with ena select
            dout <= din(1) when '0' ,
                  '0' when others ;
    end if ;
end process ;
end arc_mixedmux ;
```



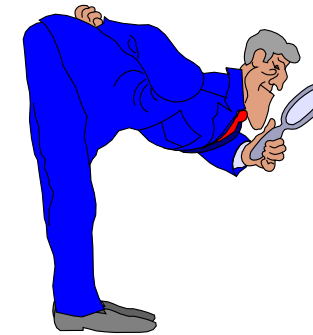
More convenient  
combinatorial  
descriptions



## Complete sensitivity list in comb system

```
-- complete sensitivity list for VHDL-2008
architecture arc_comrule2 of comrule2 is
begin
  process ( all ) -- very elegant solution !
  begin
    y <= a or b or c ;
  end process ;
end arc_comrule2 ;
```

```
-- good for latches too !
```

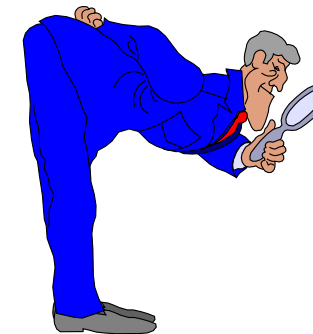


More convenient  
sync descriptions



## using rising\_edge(\_) with bit too, no dilemma

```
-- using the rising_edge(_) function with bit too
entity dff7 is
  port ( clk : in bit ;
        d   : in bit ;
        q   : out bit ) ;
end dff7 ;
architecture arc_dff7 of dff7 is
begin
  process ( clk )
  begin
    if rising_edge(clk) then -- no more dilemma !
      q <= d ;
    end if ;
  end process ;
end arc_dff7 ;
```



Improved I/O operations  
with the 'image(\_) &  
'value(\_) attributes  
(VHDL-93)



## The 'image()' attribute - VHDL-93

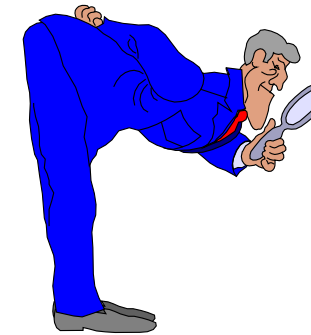
- Returns a string representing the value  
`data_type_name' image(data_value)`
- Usable only with scalar data types (not vectors)
- Useful for simulation or Verification (send strings to console)
- Here is an example

```
process ( din ) -- din is integer
begin
  assert false
  report " [time=" & time'image(now) & "]" &
    " [din=" & integer'image(din) & "]"
  severity note ;
end process ;
```

### ■ results

Note: [time=100 ns] [din=0]

Note: [time=200 ns] [din=3]



New `to_string()`  
functions



## The `to_string()` function - VHDL-2008

- Here is an example

```
process ( din ) -- din is std_logic_vector(11 downto 0)
begin
  assert false
  report " [time=" & to_string(now) & "]" &
    " [din=" & to_string(din) & "]"
  severity note ;
end process ;
```

### ■ results

Note: [time=100 ns] [din=101011001111]

Note: [time=200 ns] [din=00001111HLXZ]

- All other old built in data types are supported too
- Data types `ufixed`, `sfixed` and `float` are supported too



## `to_string()` function with more parameters

- Shown: units to use with time
- Shown: width of the right side of decimal point with real

```
process ( din ) -- din is now real
begin
  assert false
  report " [time=" & to_string(now,ps) & "]" &
    " [din=" & to_string(din,10) & "]"
  severity note ;
end process ;
```

### ■ results

Note: [time=100000 ps] [din=1.0000000000]

Note: [time=200000 ps] [din=6.5000000000]

- With type `real` you can also use other format strings like in `printf` of C:  
E%, e%, f%, g%



## to\_hstring(\_) & to\_ostring(\_) functions

- Special functions to be used on vector types:

`to_hstring( vector_type )`

`to_ostring( vector_type )`

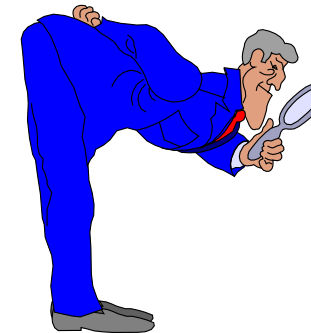
- Example with binary hex & octal presentation shown:

```
process ( din ) -- din is std_logic_vector(11 downto 0)
begin
  assert false
  report " [bin=" & to_string(din) & "]" &
        " [hex=" & to_hstring(din) & "]" &
        " [oct=" & to_ostring(din) & "]"
  severity note ;
end process ;
```

- results on "101011001111" and "00001111HLXZ"

Note: [bin=101011001111] [hex=ACF] [oct=5317]

Note: [bin=00001111HLXZ] [hex=0FX] [oct=03XX]



expressions written  
to ports when  
instantiation of  
components



## wiring of signals constants & expressions

- in VHDL-87 **only clean signals** are possible

```
ul: or_gate port map ( s , q , x1 ) ;
```

- In VHDL-93 wiring of **constants** is possible

```
ul: and_gate8
  port map(output => y
            input(3 downto 0) => x(3 downto 0)
            input(7 downto 4) => '1'&'1'&'1'&'1' ) ;
```

- in VHDL-2008 **expressions** are possible

```
ul: or_gate port map ( not s , q , x1 ) ;
```



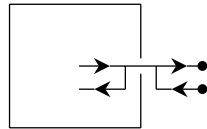
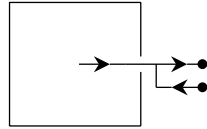
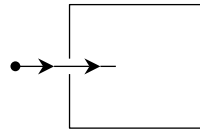
Revision of Modes  
(port directions)





## Old - Modes (port directions)

- in
  - The default mode
  - In signal can not be written (LHS) internally
  - In signal can only be read (RHS) internally
- out
  - Out signal can be externally bussed
  - Out signal can not be read back (RHS) internally !
  - Out signal can Only be written (LHS) internally
- inout
  - inout signal can be externally bussed
  - inout signal can be read back (RHS) internally
  - No internal restrictions (read write)



8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

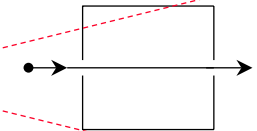
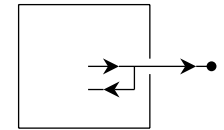
CH 07

Page: 97



## Old - Modes (port directions)

- buffer
  - Out signal can not be externally bussed
  - Out signal can be written (LHS) internally
  - Out signal can also be read (RHS) internally
  - Should be used more often !
- linkage
  - Signal passes through
  - Signal can not be read from
  - Signal can not be written
  - Rarely used (usually for power supplies)



VHDL-2002  
recommends  
removing this  
mode

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

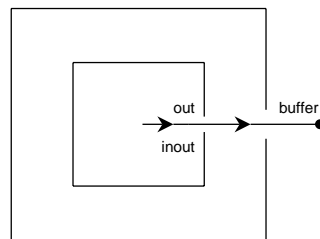
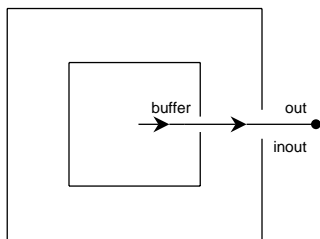
CH 07

Page: 98



## Old - Modes (port directions)

- buffer connected to out or inout from within a hierarchy
  - Warning: Can not associate buffer signal with out or inout signal
  - Restriction removed already on VHDL-2002
  - Many people did not know about it
  - Many tools ignored the 2002 revision



8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

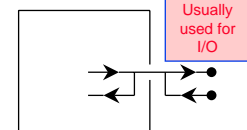
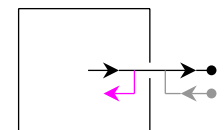
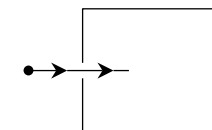
CH 07

Page: 99



## New - Modes (port directions)

- in
  - The default mode
  - In signal can not be written (LHS) internally
  - In signal can only be read (RHS) internally
- out
  - Out signal can be externally bussed
  - Out signal can be read back (RHS) internally !
  - Out signal can be written (LHS) internally
- inout
  - inout signal can be externally bussed
  - inout signal can be read back (RHS) internally
  - No internal restrictions (read write)



Will be used as normal output (not I/O) BUS

Usually  
used for  
I/O

Will be used as I/O BUS (synthesis tools prefer this mode)

People will probably stop using buffer

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 07

Page: 100



## Old coding style

```
library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_unsigned.all ;
entity count16 is
    port ( clk : in std_logic ;
          count : out std_logic_vector(3 downto 0) ) ;
end count16 ;
architecture arc_count16 of count16 is
    signal cnt : std_logic_vector(3 downto 0) ;
begin
    process (clk)
    begin
        if rising_edge(clk) then
            cnt <= cnt + 1 ;
        end if ;
    end process ;
    count <= cnt ;
end arc_count16 ;
```

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 07

Page: 101



## New coding style

```
library ieee ;
use ieee.std_logic_1164.all ;
use ieee.std_logic_unsigned.all ;
entity count16 is
    port ( clk : in std_logic ;
          count : out std_logic_vector(3 downto 0) ) ;
end count16 ;
architecture arc_count16 of count16 is
    -- signal cnt : std_logic_vector(3 downto 0) ;
begin
    process (clk)
    begin
        if rising_edge(clk) then
            count <= count + 1 ; -- now you can read mode out
        end if ;
    end process ;
    -- count <= cnt ;
end arc_count16 ;
```

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 07

Page: 102



More convenient &  
powerful generate  
statements

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 08

Page: 103



## More powerful if-generate statements

```
-- old coding style
label1: if condition
generate
    -- first alternative
    ...
end generate ;

label2: if not condition
generate
    -- second alternative
    ...
end generate ;
```

```
-- same thing in new style
label: if condition generate
    -- first alternative
    ...
else generate
    -- second alternative
    ...
end generate ;
```



8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 08

Page: 104



## Full if-generate and case-generate syntax

```
-- syntax similar to
-- regular if clause
label: if condition1
  generate
    -- first alternative
    ...
  elsif condition2 generate
    -- second alternative
    ...
  else generate
    -- last default
    alternative
    ...
  end generate ;
```

```
-- syntax similar to
-- regular case clauses
label: case
  tested_expression
  generate
    when choice1 =>
      -- first alternative
      ...
    when choice2 =>
      -- second alternative
      ...
    when others =>
      -- last default
      alternative
      ...
  end generate ;
```



## Example of case-generate statement

```
-- constant version_number : integer := 1 ;
constant version_number : integer := 2 ;
...
...
begin
  gencase: case version_number generate
    when 1 =>
      U1: comp1 port map( resetN , clk, din , dout ) ;
    when 2 =>
      U2: comp2 port map( resetN , clk, din , dout ) ;
    when others =>
      assert false
        report "Incorrect version number"
        severity error ;
  end generate ;
```



Monitoring signals from lower hierarchies using Modelsim's `signal_spy` utility (pre VHDL-2008)



## The `signal_spy` utility

- The Modelsim `signal_spy` utility can mirror signal values from a lower hierarchy (source\_object) into an existing signal in the current hierarchy (destination\_object)
- It works only in Modelsim versions later than version 5.5
- Activation is done through calling a procedure - once (see example later)
- Syntax is  
`init_signal_spy (src_object, dest_object [,verbose] ) ;`
- `src_object` and `dest_object` are strings that represent signal names. Whole signals should be used (no slices or vector elements)
- The verbose value is an optional integer and its values can be: either 0 (default) or 1
- You also need to make the following library & package visible:

```
library modelsim_lib ;
use modelsim_lib.util.all;
```



## Example of using the `signal_spy` utility

```
-- monitoring a signal from a lower hierarchy
library ieee ;
use ieee.std_logic_1164.all ;
library modelsim_lib ;
use modelsim_lib.util.all ;
entity tbspy is
  -- no ports in test bench
end tbspy ;
architecture arc_tbspy of tbspy is
  component counter
    port ( resetN , clk : in      std_logic ;
          tc       : buffer std_logic ) ;
  end component ;
  signal resetN, clk, tc : std_logic ;
  signal cnt_des : std_logic_vector(7 downto 0) ;
begin
```

making library "modelsim\_lib" & package "util" visible

A signal named "count" is an internal signal to this counter component

Destination object



## Example of using the `signal_spy` utility

```
begin
  eut: counter port map ( resetN , clk , tc ) ;
  process
  begin
    clk <= '0' ; wait for 50 ns ;
    clk <= '1' ; wait for 50 ns ;
  end process ;
  resetN <= '0' , '1' after 100 ns ;
  -- a spy process recognized by modelsim (5.5 < ver)
  process
  begin
    -- "eut/count" is ok too
    init_signal_spy("/tb/eut/count", "/cnt_des") ;
    wait ;
  end process ;
end arc_tbspy ;
```

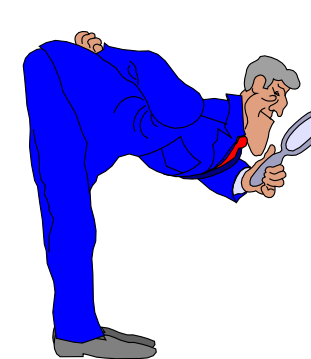
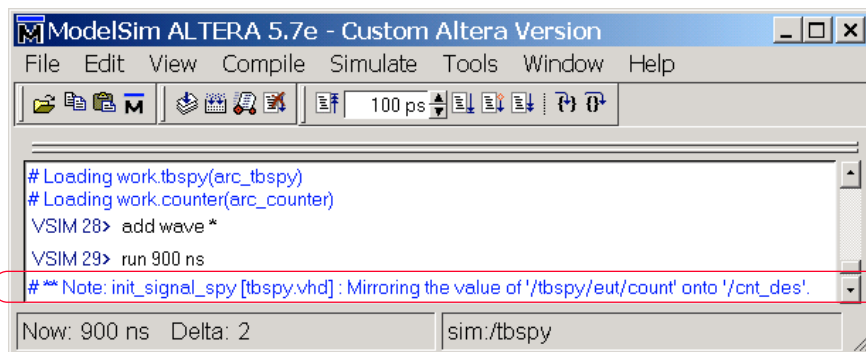
path of spy object

Destination object



## Outcome of a `verbose` activation

```
init_signal_spy( "/tb/eut/count", "/cnt_des", 1 ) ;
```



Monitoring (Accessing) signals through hierarchies using external names

Probing



## Accessing external names

- Access data objects (signals, constants & shared variables) directly and not through ports
- Full syntax:  

```
<< object_class_type pathname : subtype_indicator >>
```
- *object\_class\_type* can be: signal, constant or shared variable. *pathname* can be: absolute, relative or start from package. *subtype\_indicator*'s role is the same as it's role in alias declarations (can also be unconstrained)

- Example #1 (Absolute path)

```
ext_sig <= <<signal .tb.top_u.bot_u.int_sig : bit>> ;
```

- Example #2 (relative path - used in an assert statement):

```
assert << signal u_top.u_mid.u_bottom.count :  
          std_logic_vector(3 downto 0) >> /= "1111"  
report "count is not legal" ;
```



## Accessing external names (using alias)

- Special path characters are: ^ - go up from current hierarchy ,  
@ - start absolute hierarchy from package
- Example #3 (package with hierarchy & start from package)

```
alias data_width is << constant @work.cpu_pack.alu_pack.width : natural >> ;  
signal data_bus : std_logic_vector( data_width -1 downto 0) ;
```

- Example #4 using the **alias** command to make a compact name

```
alias count is <<signal .eut.u_top.cnt : std_logic_vector>> ;  
assert count(3 downto 0) = "1001"  
report "got: "& to_string(count) & "/= 1001 expected" ;
```

- Results of Example #4 (when count is not 1001)

Error: got: 0000 /= 1001 expected



## Force & Release assignments

- Override all normal resolved values assigned to a signal (similar to script force command)
  - Initialize signals quickly bypassing complex initialization sequences
  - Inject erroneous values into design to ensure that detection & response are appropriate

- A verification feature

- Force Syntax:

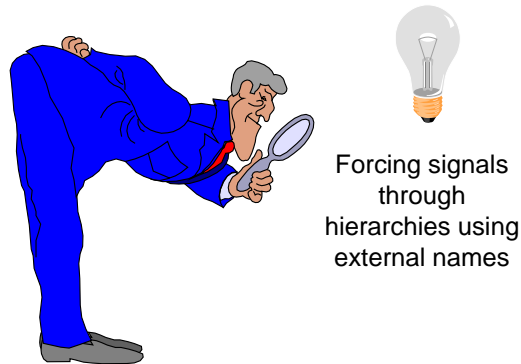
```
signal_name <= force expression ;
```

*signal\_name* can be a regular signal name but also can also be (and usually is) an external name or alias

- Release syntax:

```
signal_name <= release ;
```

- Stop forcing & design takes back control



## Force & Release assignment example

```

process
  use work.control_pkg.all ;
  alias clk is << signal eut.clk : std_logic >> ;
  alias present_state is
    << signal eut.sm.present_state :
      std_logic_vector(3 downto 0) >>;
begin
  . . .
  -- force a bad (corrupt) state to state machine
  wait until falling_edge(clk) ;
  current_state <= force "1111" ; -- bad (corrupt) state
  wait until falling_edge(clk) ;
  current_state <= release ;
  -- monitor recovery activity
  . . .
end process ;

```

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 08

Page: 117



Forcing ports:  
Driving value &  
Effective value

8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

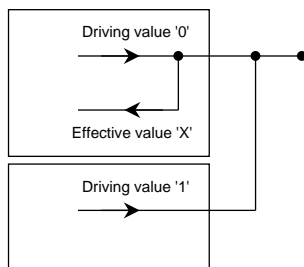
CH 08

Page: 118



## Force & Release assignments of ports

- Driving value
  - The value presented externally by the entity
  - Determined by internal sources within the entity
- Effective value
  - The value seen internally by an entity
  - Determined by whatever is externally connected to port



VHDL-2008 allows us to force the driving and effective values of a port independently!



8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

CH 08

Page: 119



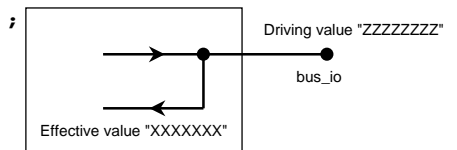
## Force & Release assignments of ports

- VHDL-2008 allows us to force the driving and effective values of a port independently !
- Forcing driving value (can be done to any port of mode **out** or **inout** or **buffer** but can not be done to port of mode **in**)
 

```
bus_io <= force out "ZZZZZZZZ" ;
```
- Forcing effective value (can be done to port of any mode)
 

```
bus_io <= force in "XXXXXXXX" ;
```
- Releasing previous ports
 

```
bus_io <= release out ;
bus_io <= release in ;
```
- Default forcing assignment of port signals of mode **in** is **force in** (forcing effective value)
- Default forcing assignment of port signals of mode **inout,out,buffer** is **force out** (forcing driving value)



8/5/2012

VHDL(13) - Amos Zaslavsky © Copyright

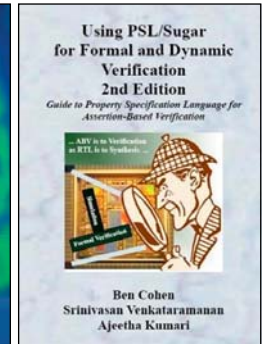
CH 08

Page: 120



## General Information on PSL

- PSL = Property Specification Language
- Origin: Sugar (IBM) => IEEE std 1850
- Allows specification of temporal properties of a model
  - Enables Static Verification (using formal proof tools)
  - Enables Dynamic Verification (using simulation checkers)
- For full information refer to:
  - *"Practical Introduction to PSL"*, Cindy Eisner & Dana Fisman
  - *"Using PSL/Sugar for Formal and Dynamic Verification"*, multiple authors



Embedded PSL



## More advanced topics

- IP (Intellectual Property) protection
  - Pragma based approach
  - Allows IP authors to mark specific areas of VHDL code for encryption
  - using standard cipher algorithms (symmetric & asymmetric ciphers such as: DES, AES, blowfish, twofish, serpent, cast, RSA, elgama, and others)
  - using standard encoding methods (such as: uudecode, base64 & others)
  - Using standard hash functions (sha1, md5, md2, ripemod-160)
  - Support is embedded in standard simulation & Synthesis tools
- VHDL Procedural Interface (VHPI)
  - application-programming interface (API) to VHDL tools
  - Tools can gain access about VHDL model during analysis, elaboration or execution of model
  - Can help development of add-in tools such as: **Linters**, **Code coverage analyzer tools**, **Timing and Power analyzers**
  - Can help usage of external simulation models



More advanced features



# Open Source VHDL Verification Methodology



## Why OS-VVM?



Standard VHDL has all the features necessary to code randomization of stimulus and functional coverage – both very important while verifying larger, system-level designs. The problem is that those features are quite advanced and require high coding skills. That's why Open Source VHDL Verification Methodology is so important: it creates a couple of easily accessible VHDL packages that hide quite arcane implementation details from the average user, making generation of random stimulus and intelligent functional coverage not only easy, but also pleasurable task.

## Benefits of OS-VVM

- Based on VHDL 2008, can work with VHDL 2002
- Provides advanced capabilities for random value generation and functional coverage
- Randomizes values with uniform or weighted distributions
- Also supports favor\_small, favor\_big, Gaussian and Poisson distribution
- Works perfectly with Transaction Level Modeling
- Enables intelligent randomization based on the functional coverage holes (bins that are not covered)
- Allows definition of normal, illegal and ignore bins for regular coverage and cross-coverage
- Equipped with flexible coverage reporting procedures



Please [create an account](#) to get started.

Username

Password

Remember Me

### Most recent blog posts

[Webinar about OS-VVM coming soon. Any questions?](#)  
[OS-VVM Updates](#)  
[Let's Meet At DAC](#)  
[RandCoPoint for ItemPoint coverage](#)  
[randc in OSVVM – another view](#)



# Coffee Break

